FREQUENCY DOMAIN ANALYSIS OF DYNAMIC SYSTEMS

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Norms

• Consider a linear continuous time invariant system with transfer function

$$H(s) = C(sI - A)^{-1}B + D \in \mathbb{C}^{r \times m}$$

where $j\omega \in \mathcal{D}(H)$ for all $\omega \in \mathbb{R}$. The impulse response is

$$h(t) = Ce^{At}B + D\delta(t) \in \mathbb{R}^{r \times m}$$

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Norms based on either time response h(t) or frequency response $H(j\omega)$



• Using the Parseval's theorem it is verified that

Fact

The following equality holds

$$\int_0^\infty \operatorname{Tr}(h(t)'h(t)) \ dt = \frac{1}{\pi} \int_0^\infty \operatorname{Tr}(H(j\omega)^\sim H(j\omega)) d\omega$$

- It is important to notice that :
 - the above equality still holds if h(t) and H(s) are replaced by h(t)' and H(s)' respectively. That is

 $(A, B, C, D) \Longrightarrow (A', C', B', D')$

H_2 norm

• Let S be a linear time invariant system with transfer function (or state space representation) defined by matrices of compatible dimensions, denoted as S = (A, B, C, D).

Lemma (H_2 norm)

The H_2 norm of system S is given by

$$|\mathcal{S}||_{2}^{2} := \int_{0}^{\infty} \operatorname{Tr}(h(t)'h(t)) \ dt = \frac{1}{\pi} \int_{0}^{\infty} \operatorname{Tr}(H(j\omega)^{\sim}H(j\omega)) d\omega$$

 $\bullet\,$ The above equalities hold whenever ${\mathcal S}$ is asymptotically stable, that is

$$j\omega \in \mathcal{D}(H)$$
, $\forall \omega \in \mathbb{R}$

Norms

H_2 norm

• Important : $\|S\|_2 < \infty$ if and only if D = 0. The H_2 norm is finite only for strictly proper systems. From the previous Lemma we have

$$\|\mathcal{S}\|_{2}^{2} = \int_{0}^{\infty} \operatorname{Tr}\left(B'e^{A't}C'Ce^{At}B\right)dt + \operatorname{Tr}(B'C'D) + \operatorname{Tr}(D'CB) + \operatorname{Tr}(D'D)\int_{0}^{\infty}\delta(t)^{2}dt$$

• From Parseval's theorem :

$$\int_0^\infty \delta(t)^2 dt = \frac{1}{\pi} \int_0^\infty d\omega = +\infty$$

Then, $\|S\|_2$ finite requires D = 0.



• For strictly proper time invariant systems the H₂ norm is calculated as follows :

Lemma (H_2 norm calculation)

The following hold :

• $\|S\|_2^2 = \operatorname{Tr}(B'P_oB)$ where P_o is the observability gramian :

$$P_o = \int_0^\infty e^{A't} C' C e^{At} dt$$

• $\|S\|_2^2 = \operatorname{Tr}(CP_cC')$ where P_c is the controllability gramian :

$$P_c = \int_0^\infty e^{At} BB' e^{A't} dt$$

H_2 norm

- Other possibilities for H_2 norm calculation using LMI :
 - Bounding the observability gramian

$$\begin{split} \mathcal{S}\|_{2}^{2} &= \inf_{X>0} \left\{ \mathrm{Tr}(B'XB) : A'X + XA + C'C < 0 \right\} \\ &= \sup_{Y>0} \left\{ \mathrm{Tr}(B'YB) : A'Y + YA + C'C > 0 \right\} \end{split}$$

• Bounding the controllability gramian

$$\begin{split} \|S\|_{2}^{2} &= \inf_{X>0} \{ \operatorname{Tr}(CXC') : AX + XA' + BB' < 0 \} \\ &= \sup_{Y>0} \{ \operatorname{Tr}(CYC') : AY + YA' + BB' > 0 \} \end{split}$$

Using LMI solvers, all these problems provide the H_2 norm of the system S within a precision defined by the designer.

H_{∞} norm

• Let $V \in \mathbb{C}^{r imes m}$ be a complex matrix. The matrix

$$Q = V^{\sim}V \in \mathbb{C}^{m \times m}$$

is Hermitian and positive semidefinite, that is :

- $Q^{\sim} = Q$ and $v^{\sim}Qv \ge 0$ for all $v \in \mathbb{C}^m$.
- The eigenvalues of Q satisfies $\lambda_i(Q) \ge 0$ for all $i = 1, \dots, m$.
- The quantities

$$\sigma_i(V) := \sqrt{\lambda_i(Q)} = \sqrt{\lambda_i(V^{\sim}V)}, \quad i = 1, \cdots, m$$

are the singular values of V.

The quantity

$$\|V\|_{\infty} := \max_{i=1,\cdots,m} \sigma_i(V) := \sigma_M(V)$$

is the ∞ -norm of V. Moreover $||V||_{\infty} = ||V'||_{\infty}$.

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• Let S be a linear time invariant system with transfer function (or state space representation) defined by matrices of compatible dimensions, denoted as S = (A, B, C, D).

Lemma (H_{∞} norm)

The H_∞ norm of system ${\mathcal S}$ is given by

$$\|\mathcal{S}\|_{\infty} := \sup_{\omega \in \mathbb{R}} \sigma_{M}(H(j\omega))$$

 $\bullet\,$ The above equality holds whenever ${\mathcal S}$ is asymptotically stable, that is

$$j\omega \in \mathcal{D}(H)$$
, $\forall \omega \in \mathbb{R}$

H_{∞} norm

• Time domain interpretation : Consider the state space representation of H(s) as

$$\dot{x}(t) = Ax(t) + Bw(t), x(0) = 0$$

 $z(t) = Cx(t) + Dw(t)$

where $0 \in \mathcal{D}(\hat{w})$. The output $\hat{z}(s) = H(s)\hat{w}(s)$ gives

$$\int_{0}^{\infty} z(t)'z(t)dt = \frac{1}{\pi} \int_{0}^{\infty} \underbrace{\hat{w}(j\omega)^{\sim} H(j\omega)^{\sim} H(j\omega) \hat{w}(j\omega)}_{\hat{z}(j\omega)^{\sim} \hat{z}(j\omega)} d\omega$$
$$\leq \|\mathcal{S}\|_{\infty}^{2} \int_{0}^{\infty} w(t)'w(t)dt$$

 $\mathsf{Fact} : \|\mathcal{S}\|_{\infty} \leq \gamma \Longleftrightarrow \|z(t)\|_2 \leq \gamma \|w(t)\|_2$

H_{∞} norm

• Using the quadratic Lyapunov function v(x) = x'Px with P > 0 and imposing

$$\dot{w}(x(t))\leq -z(t)'z(t)+\gamma^2w(t)'w(t)\;,\;\;orall t\geq 0$$

for some $\gamma \geq 0$, the time integration from 0 to $+\infty$ provides

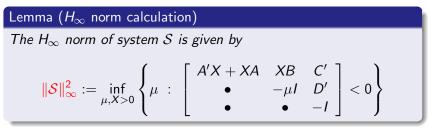
$$\int_0^\infty z(t)'z(t)dt - \gamma^2 \int_0^\infty w(t)'w(t)dt \le 0$$

yielding the conclusion that $\|S\|_{\infty} \leq \gamma$. On the other hand, taking into account the state space representation of S we get

$$\dot{v}(x(t)) = (Ax(t) + Bw(t))'Px(t) + x(t)'P(Ax(t) + Bw(t))$$

H_∞ norm

• For a linear time invariant system with transfer function (or state space representation) defined by matrices of compatible dimensions, denoted as S = (A, B, C, D), the H_{∞} norm is calculated as follows :



This problem can be solved with no big difficulty since it is expressed by an LMI with respect to the variables $\mu \in \mathbb{R}$ and $X = X' \in \mathbb{R}^{n \times n}$.

KYP Lemma

KYP Lemma

 Is one of the most general results on frequency domain. Consider a transfer function H(s) with state space representation S = {A, B, C, D}, det(jωI - A) ≠ 0, ∀ ω ∈ ℝ and a symmetric matrix Π of compatible dimension.

Lemma (KYP Lemma)

The transfer function H(s) satisfies the constraint

$$\left[\begin{array}{c}I\\H(j\omega)\end{array}\right]^{\sim}\Pi\left[\begin{array}{c}I\\H(j\omega)\end{array}\right]<0 \ \forall \omega\in\mathbb{R}$$

if and only if there exists P = P' such that

$$\begin{bmatrix} A'P + PA & PB \\ B'P & 0 \end{bmatrix} + \begin{bmatrix} 0 & I \\ C & D \end{bmatrix}' \Pi \begin{bmatrix} 0 & I \\ C & D \end{bmatrix} < 0$$

KYP Lemma

KYP Lemma

• Sufficiency is simple to prove from the matrix function

$$\mathcal{A}(s) := (sI - A)^{-1}$$

which allows us to verify that :

- AA(s) = -I + sA(s) for all $s \in \mathbb{C}$.
- For any $P = P' \in \mathbb{R}^{n \times n}$ the matrix function

 $\mathcal{Q}(\omega) := \mathcal{A}(j\omega)^{\sim} (\mathcal{A}'\mathcal{P} + \mathcal{P}\mathcal{A})\mathcal{A}(j\omega) + \mathcal{A}(j\omega)^{\sim}\mathcal{P} + \mathcal{P}\mathcal{A}(j\omega)$

satisfies $\mathcal{Q}(\omega) = 0$ for all $\omega \in \mathbb{R}$.

• The following factorization of H(s) holds

$$\left[\begin{array}{cc} 0 & I \\ C & D \end{array}\right] \left[\begin{array}{cc} \mathcal{A}(s)B \\ I \end{array}\right] = \left[\begin{array}{cc} I \\ H(s) \end{array}\right], \ \forall s \in \mathbb{C}$$

KYP Lemma

KYP Lemma

 Multiplying the second inequality of the KYP Lemma to the left by [B'A(jw)~ I] and to the right by its transpose, from the previous results we obtain

$$B'\mathcal{Q}(\omega)B + \left[\begin{array}{c}I\\H(j\omega)\end{array}\right]^{\sim} \Pi \left[\begin{array}{c}I\\H(j\omega)\end{array}\right] < 0 \ \forall \omega \in \mathbb{R}$$

and the first inequality of the KYP Lemma holds due to the fact that $\mathcal{Q}(\omega) = 0$ for all $\omega \in \mathbb{R}$.

- The necessity states that if the first inequality of the KYP Lemma holds for some matrix Π then the second one also holds for the same Π for some matrix P = P'.
- If A is asymptotically stable then P = P' > 0 can be included with no loss of generality whenever $C'\Pi_{22}C \ge 0$.

 H_{∞} theory

Comparison

• It is seen that

$$\|H(s)\|_{\infty} < \gamma \iff \Pi = \begin{bmatrix} -\gamma^2 I & 0 \\ 0 & I \end{bmatrix}$$

The celebrated Small Gain Theorem is a mere particular case of the KYP Lemma which provides

$$\|H(s)\|_{\infty} < \gamma \iff \begin{bmatrix} A'P + PA + C'C & PB + C'D \\ B'P + D'C & D'D - \gamma^{2}I \end{bmatrix} < 0$$

for some P > 0. The Schur Complement gives the previous LMI for H_{∞} calculation with $\mu = \gamma^2$.

Locus in the frequency domain

Locus in the *s*-space

 The is no difficulty to impose several constraints on H(jω) characterized by different matrices Π_i, i = 1, · · · , N. From the KYP Lemma the inequalities

$$\begin{bmatrix} I\\ H(j\omega) \end{bmatrix}^{\sim} \Pi_{i} \begin{bmatrix} I\\ H(j\omega) \end{bmatrix} < 0 \quad \forall \omega \in \mathbb{R}$$

holds for all $i = 1, \dots, N$ if and only if there exist matrices $P_i = P'_i$ such that

$$\begin{bmatrix} A'P_i + P_iA & P_iB \\ B'P_i & 0 \end{bmatrix} + \begin{bmatrix} 0 & I \\ C & D \end{bmatrix}' \prod_i \begin{bmatrix} 0 & I \\ C & D \end{bmatrix} < 0$$

holds for all $i = 1, \cdots, N$.

Locus in the frequency domain

Locus in the *s*-space

- Examples for SISO systems : $H(j\omega) \in \mathbb{C}$ for each $\omega \in \mathbb{R}$
 - Circle : $|H(j\omega)| < \gamma$ for all $\omega \in \mathbb{R}$

$$H(j\omega)^*H(j\omega) < \gamma^2 \iff \Pi = \begin{bmatrix} -\gamma^2 & 0\\ 0 & 1 \end{bmatrix}$$

• Right hand side of C : $\operatorname{Re}(H(j\omega)) > 0$ for all $\omega \in \mathbb{R}$

$$\Pi = \left[\begin{array}{cc} 0 & -1 \\ -1 & 0 \end{array} \right]$$

• Strip : $\alpha < \operatorname{Re}(H(j\omega)) < \beta$ for all $\omega \in \mathbb{R}$

$$\Pi_1 = \left[\begin{array}{cc} -2\beta & 1 \\ 1 & 0 \end{array} \right] \ , \ \Pi_2 = \left[\begin{array}{cc} 2\alpha & -1 \\ -1 & 0 \end{array} \right]$$

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Control design using LMIs

State feedback

• Consider the linear time invariant system :

$$\dot{x}(t) = Ax(t) + B_1w(t) + B_2u(t) z(t) = Cx(t) + D_1w(t) + D_2u(t) u(t) = Kx(t)$$

The closed loop transfer function is given by

$$H(s) = (C + D_2 \mathbf{K})(sI - (A + B_2 \mathbf{K}))^{-1}B_1 + D_1$$

The determination of matrix K using LMIs is related to the transfer function G(s) = H(s)', given by

$$G(s) = B'_1(sI - (A + B_2K)')^{-1}(C + D_2K)' + D'_1$$

Control design using LMIs

State feedback

• Given matrices Π_1, \cdots, Π_N , the inequalities

$$\begin{bmatrix} I \\ G(j\omega) \end{bmatrix}^{\sim} \Pi_{i} \begin{bmatrix} I \\ G(j\omega) \end{bmatrix} < 0 \quad \forall \omega \in \mathbb{R}$$

for $i = 1, \cdots, N$ are satisfied if there exists a matrix P = P' such that

$$\begin{bmatrix} (A + B_2 \mathbf{K})P + P(A + B_2 \mathbf{K})' & P(C + D_2 \mathbf{K})' \\ (C + D_2 \mathbf{K})P & 0 \end{bmatrix} + \\ + \begin{bmatrix} 0 & I \\ B'_1 & D'_1 \end{bmatrix}' \Pi_i \begin{bmatrix} 0 & I \\ B'_1 & D'_1 \end{bmatrix} < 0$$

holds for all $i = 1, \dots, N$.

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Control design using LMIs

State feedback

 \bullet Assuming that all matrices Π_1,\cdots,Π_N are such that

$$\begin{bmatrix} 0 & B_1 \end{bmatrix} \prod_i \begin{bmatrix} 0 \\ B'_1 \end{bmatrix} \ge 0 , \ i = 1, \cdots, N$$

the asymptotical stability of the closed loop system $A + B_2 K$ is assured by the linear constraint P > 0.

• The previous inequalities expressed in terms of the matrix variables (P > 0, K) are converted into LMIs with respect to the matrix variables (P > 0, L) where

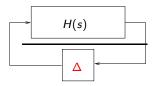
$K = LP^{-1}$

Since *P* can not depend on the index $i = 1, \dots, N$ the necessity part of the KYP Lemma is lost. For N = 1 the necessity obviously holds.

Stability

Stability

• Systems with special structure - Linear part plus feedback



State space realization

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bw(t) \\ z(t) &= Cx(t) + Dw(t) \\ w(t) &= \Delta z(t) , \ \Delta \in \Xi \end{aligned}$$

 \equiv includes *a priori* information about the structure of \triangle

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Stability

Stability

Assuming that

$$\left[\begin{array}{c} \Delta \\ I \end{array}\right]' \Pi \left[\begin{array}{c} \Delta \\ I \end{array}\right] \ge 0 \ , \ \forall \ \Delta \in \Xi$$

then multiplying to the left by z' and to the right by z and using the fact that $w = \Delta z$ we conclude that

$$\left[\begin{array}{c}w\\z\end{array}\right]'\Pi\left[\begin{array}{c}w\\z\end{array}\right]\geq 0$$

for all (w, z) of appropriate dimensions such that $w = \Delta z$. • On the other hand, considering the Lyapunov function

$$v(x) = x' P x , \quad P = P' > 0$$

Stability

Stability

and imposing that its time derivative along any trajectory of the system under consideration satisfies

$$\dot{v}(x) + \left[\begin{array}{c} w \\ z \end{array} \right]' \Pi \left[\begin{array}{c} w \\ z \end{array} \right] < 0$$

asymptotic stability is preserved for all $\Delta\in\Xi.$ The key point on this algebraic manipulation is that the equality

$$\left[\begin{array}{c} w\\ z\end{array}\right] = \left[\begin{array}{cc} 0 & I\\ C & D\end{array}\right] \left[\begin{array}{c} x\\ w\end{array}\right]$$

and the KYP Lemma provide the next important result.

Stability

Stability

Lemma (Stability)

Assume that $0 \in \Xi$. The previous feedback structure is stable for all $\Delta \in \Xi$ provided there exists a symmetric matrix Π such that

$$\begin{bmatrix} I\\ H(j\omega) \end{bmatrix}^{\sim} \Pi \begin{bmatrix} I\\ H(j\omega) \end{bmatrix} < 0 \quad \forall \omega \in \mathbb{R}$$

and

$$\left[\begin{array}{c} \Delta \\ I \end{array}\right]' \Pi \left[\begin{array}{c} \Delta \\ I \end{array}\right] \ge 0 \ , \ \forall \ \Delta \in \Xi$$

- The assumption 0 ∈ Ξ is necessary since P = P' must be positive definite in order to be used in the Lyapunov function.
- Both constraints are convex (linear) with respect to Π .

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Stability

Stability

- Important aspects: Some situations can be dealt with no particular difficulty :
 - The classical H_{∞} stability condition is obtained if Π is imposed to be

$$\Pi = \left[\begin{array}{cc} -\gamma^2 I & 0 \\ 0 & I \end{array} \right]$$

for some $\gamma >$ 0. Asymptotic stability is preserved whenever

$$\|H(s)\|_{\infty} < \gamma$$
, $\|\Delta\|_{\infty} \le \gamma^{-1}$

- Matrix Π can be parameter dependent, that is $\Pi = \Pi(\Delta)$.
- From the KYP Lemma, matrix P can also be parameter dependent, that is $P = P(\Delta)$.
- For polytopic systems with Ξ = co{Δ₁, · · · , Δ_N} matrices (P_i > 0, Π_i) can be used.

Problems

Problems

1. Consider the following asymptotically stable transfer functions

•
$$H(s) = \frac{(s+2)}{(s^2+2s+2)(s+1)}$$

• $H(s) = \frac{(s-2)}{(s^2+2s+2)(s+1)}$
• $H(s) = \frac{(s-2)^2}{(s^2+2s+2)(s+1)}$

Determine the H_2 norm of each transfer function using gramians and a numerical routine of LMIsolver.

2. Consider the following asymptotically stable transfer functions

•
$$H(s) = \frac{(s+2)}{(s^2+2s+2)(s+1)}$$

• $H(s) = \frac{(s-2)}{(s^2+2s+2)(s+1)}$
• $H(s) = 1 + \frac{(s-2)^2}{(s^2+2s+2)(s+1)}$

Determine the H_{∞} norm of each transfer function using the singular value diagram and a numerical routine of LMIsolver.

Problems

Problems

3. Consider a linear plant with transfer function H(s) plus a nonlinear feedback $w(t) = \Delta(z(t))$. Show that asymptotical stability is preserved whenever

$$\|H(s)\|_{\infty} < \gamma \;,\;\; \Delta(z)'\Delta(z) \leq \gamma^{-2}z'z$$

4. Use the solution of the previous problem to verify that the origin for a SISO system defined by

$${\cal H}(s)=rac{(s^2+3s+3)(s-1)}{(s+1)^3}\,,\,\,\,\Delta(z)=\left\{egin{array}{cc} 1-e^{-z/4},&z\geq 0\ -1+e^{z/4},&z\leq 0 \end{array}
ight.$$

is globally asymptotically stable.

Problems

Problems

5. Consider a linear plant with transfer function H(s) plus a linear feedback $w(t) = -\delta z(t)$ where $\delta \in \mathbb{R}$. For H(s) given by

$$H(s) = \frac{(s-1)(s-2)}{(s+1)^2(s+4)}$$

determine the upper bound δ_{max} such that asymptotic stability is preserved for all 0 $\leq \delta \leq \delta_{max}$ using :

- The H_{∞} theory
- The Routh criterion.

Compare the obtained results from the root locus plot (with respect to δ) of the closed loop system.

Problems

Problems

6. Consider the function

$$\hat{f}(s) = \frac{(1 - e^{-sT})}{s}$$

- Is it possible to determine $\|\hat{f}(s)\|_{\infty}$? Why ?
- In the affirmative case, determine $\|\hat{f}(s)\|_{\infty}$.
- Is it possible to determine $\left\|\frac{e^{-T_s}}{s}\right\|_{\infty}$?
- 7. Consider a time delay system with characteristic equation $P(s) + \kappa e^{-Ts} = 0$ where $\kappa, T \ge 0$. Using the previous result show that asymptotic stability is preserved whenever

$$T\left\|\frac{\kappa s}{P(s)+\kappa}\right\|_{\infty} < 1$$

Problems

Problems

- 8. Consider H(s) and G(s) two asymptotically stable transfer functions. Show that the following relations hold :
 - $\|H(s)G(s)\|_2 \le \|H(s)\|_{\infty} \|G(s)\|_2$
 - $\|H(s)G(s)\|_{\infty} \le \|H(s)\|_{\infty} \|G(s)\|_{\infty}$
- 9. Verify the results of the previous problem for

$$H(s) = rac{(s+1)^4}{(s+2)^5} \;, \;\;\; G(s) = rac{(s-1)}{(s+1)^2}$$

10. Determine numerically the values of $||G(s)||_2$ and $||G(s)||_{\infty}$ for the transfer function

$$G(s) = \frac{e^{-Ts}}{s^3 + 4s^2 + 4s + \kappa e^{-Ts}}$$

defined with $\kappa = 1/2$ and T = 2.

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